APPLICATION NOTE



STEP-DOWN VOLTAGE CONVERTER

INTRODUCTION

In many cases, electronic circuits require an accurate and stable voltage supply. The voltage provided by transformers, batteries, generators, and so on are unstable and not very accurate. Therefore, a voltage regulator must be placed between the source and the load.

In the past, linear voltage regulators were used for this task. They were inexpensive, stable, exact, and very easy to use. However, they burned great quantities of fuel, though, which meant they wasted a lot of power.

The alternative solution was switching regulators. Though switching regulators were expensive, complicated, and generated ripple, they were very economical. In fact, designs of switching regulators have greatly improved in recent years. The disadvantages have declined, and switching regulators are widely used today.

ADVANTAGES

The step down voltage converter is probably the most commonly used switching regulator. Using a ZiLOG microcontroller (MCU), this application note describes a basic step down voltage converter. The step down voltage converter can drive all kinds of loads (ohmic, inductive, capacitive), including:

- Halogen light bulbs
- Motors
- Relays

The main advantage of this step down voltage converter, compared to others is that the ZiLOG converter is able to perform other tasks as well. Using this converter, an engineer can implement many designs including an IR-receiver, revolution-regulation, current surveillance, or power supply failure handler.

There is abundant literature available explaining the function of a step down converter. This application note is intended to explain one form of implementation.

REQUIREMENTS

The first requirement for creating a step down voltage converter is a Pulse Width Modulator (PWM). The MCU essentially generates the PWM. There are two MCU choices:

- The Z86E02
- The Z86E31

The Z86E02 is the most cost effective and generates PWMs up to almost 8kHz. If a 16kHz PWM is required, use the Z86E31. This 7-bit PWM has fair resolution. If the PWM frequency is raised, then the resolution must be lowered.

REQUIREMENTS (Continued)

The second requirement is a switch. A switch is made from the transistor network (BC548B + TIP116), which can source up to 4A. The design engineer must consider a heat sink for the TIP116 for this specific application. The following calculation illustrates this approximation:

 $P_{heat} = Uce * I * fpwm * (t_{on}+t_{off})$

 P_{heat} = power with which the transistor silicon is heated [W] UCE = voltage across collector and emitter [V] \hat{I} = maximum current through the transistor /load [A] fpwm = PWM frequency [Hz] t_{on}, t_{off} = transistor switching times [s]

The heat transmission resistance of the heat sink can be calculated as follows:

$$R_{th} - (\underbrace{\hat{T}_{silicon} - \hat{T}_{ambient}}_{P_{heat}})$$

$$R_{th} = \text{thermal resistance from silicon to ambient} \begin{bmatrix} \underline{K} \\ \overline{W} \end{bmatrix}$$

$$\widehat{T}_{silicon} = \text{maximum silicon temperature [K]}$$

$$\widehat{T}_{ambient} = \text{maximum ambient temperature [K]}$$

The necessary thermal resistance of the heat-sink is represented as:

$$\begin{aligned} R_{th \ heat-sink} &= R_{th} - R_{th \ transistor} + R_{th \ transistor} > heat-sink \\ R_{th \ heat-sink} &= thermal \ resistance \ from \ heat \ sink \ to \ ambient \begin{bmatrix} \frac{K}{W} \\ W \end{bmatrix} \\ R_{th \ transistor} &= thermal \ resistance \ from \ silicon \ to \ package \begin{bmatrix} \frac{K}{W} \\ W \end{bmatrix} \\ R_{th \ transistor} > heat-sink &= thermal \ resistance \ \begin{bmatrix} \frac{K}{W} \\ W \end{bmatrix} \end{aligned}$$

The thermal resistance from the transistor package to the heat sink depends on whether a thermal coat, glimmer isolation (ca. 2K/W), or crema (0.5K/W) is used.

The third part necessary for a step down converter is the inductance. The minimal inductance (L1) value is calculated with the following formula:

$$L_{1} = \frac{U_{0}}{4*f_{pwm}*\Delta I}$$

$$L_{1} = \text{inductance } [H]$$

$$U_{0} = \text{supply voltage } [V]$$

$$f_{pwm} = \text{PWM frequency } [Hz]$$

$$\Delta I = \text{maximum current ripple } [A]$$

To keep the current ripple low, high inductance values are preferable. However, to keep the regulation loop stable, low inductance values are better. Keep the inductance value as small as possible to maintain regulation loop stability.

VOLTAGE CONTROL

To make a step down voltage converter, the voltage must be controlled. This control is accomplished by feeding the voltage across the load back into the MCU. The load voltage is available at port 31, and an adjustable reference voltage is available at port 33. The MCU compares the two voltages and adjusts the duty cycle of the PWM accordingly, provided that the two voltages are not the same. When the voltage across the load is too low compared with the reference voltage, then the MCU increases the duty cycle of the PWM. When the voltage across the load is too high compared with the reference voltage, then the MCU decreases the duty cycle of the PWM. P1 adjusts the reference voltage and depending on this reference the MCU shortens or lengthens the duty cycle of the PWM. This condition causes the load voltage to lower or rise.

REGULATION LOOP

The step down voltage converter is a regulation loop. The program provides a simple integrating regulator with a fixed phase shift of 90°. The transistor network is an amplifier. The inductor is a PT1, with a maximum 90° phase shift. The voltage measuring is done by R5-R6, which is an amplifier with an amplification of 0.325. Consequently, there is a maximum phase shift of 180° with an ohmic or inductive load. Ohmic loads are neutral, whereas inductive loads decrease the stability. With a capacitive load, however, the phase shift can reach a maximum of 270°. Therefore, with an ohmic or inductive load, the step down converter can oscillate; with a capacitive load, the converter can be unstable under worst-case conditions.

To avoid this problem, keep the value of the inductance (L1) as low as possible. If swinging tendencies of the step down converter are still too high, reduce the value of *integration_time* in the program. As a final step, reduce the input voltage.

Follow these steps when the swinging tendency is too high. Under normal conditions, these steps are not necessary. Figure 1 illustrates the regulation loop sequence.

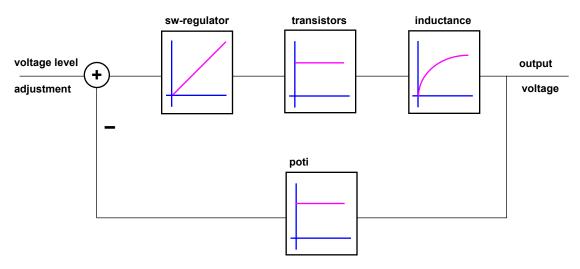


Figure 1. Regulation Loop Diagram for Ohmic or Inductive Loads

THE MAIN PROGRAM

The definition of variables, constants, and macros are defined at the beginning of the software code. The constants may be altered to customize the voltage-regulator. Next, set the *Softstart* feature to be active or inactive. From that point, select either the Z86E02 or the Z86E31 as the target controller.

The Z8 initializes and enters the main program. The main program consists of NOPs which can be replaced by custommade subroutines. The regulator and the PWM generator are interrupt-driven.

First, load the timer (T1) with the next half-period of the PWM cycle. Next, enter the regulator routine. From that point, the PWM is adjusted to achieve the pre-adjusted value of the voltage. Finally, exit the interrupt; the Z8 returns to the main

SCHEMATICS (Continued)

program. The Z8 is a very robust and fast MCU, and there is enough core performance left to execute various other tasks in the main program. Figure 2 illustrates the main program sequence. Figure 3 illustrates the interrupt routine.

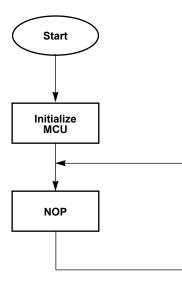


Figure 2. Program Sequence

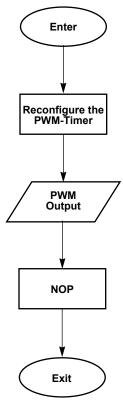


Figure 3. Interrupt Routine

SCHEMATICS

Figures 4 and 5 illustrate the schematic diagrams of the two chips.

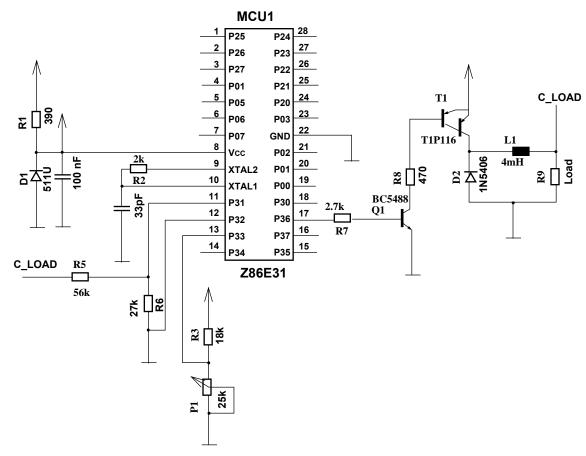


Figure 4. Voltage Regulator Schematic Using the Z86E31

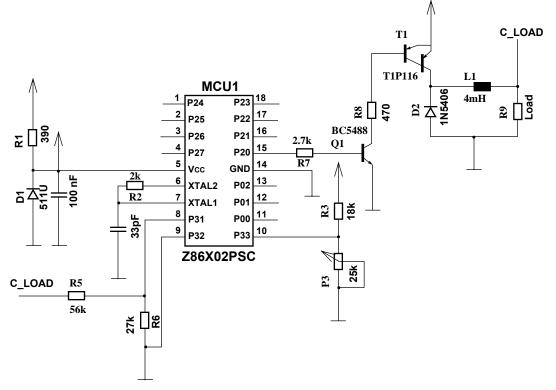


Figure 5. Voltage Regulator Schematic Using the Z86X02PSC

SAMPLE CODE (Continued)

SAMPLE CODE

Pages 7–13 illustrate the step-by-step process of creating a step down voltage regulator.

CONCLUSION

By following the basic guidelines contained in this application note, any design engineer can ensure a quick and easy implementation of a step down voltage converter for a variety of design applications.

This application note is for 8MHz. ; "Using A Z86E02 MCU For "Step Down Voltage Regulation" via pwm. ; ; ; FILE: vol_reg1.S ; DATE: 23.12.97 MCU: Z86E02 ; PROJECT: Voltage Regulator ; AUTHOR: Klaus Buchenberg ; SOFTWARE: REVISION 3.0 ; This program is assembled by ZiLOG ZMASM assembler ; ;******** GLOBALS ON When the softstart option is wished, then set SOFTSTART_WISHED to 1 = Yes SOFTSTART_WISHED.equ 0 ; 1 = Yes 0 = NoBitnumber definitions ; bitno0 .equ%01
bitno1 .equ%02 bitno2 .equ%04 bitno3 .equ%08 bitno4 .equ%10 bitno5 .equ%20 bitno6 .equ%40 bitno7 .equ%80 PORTS DEFINITION ; Port 0 pin ; P00: power. UP key ; .equ bitno0 ; not used ;upkey P01: power. DOWN key ; ;downkey.equ bitnol; not used P02: idle input ; Port 2 pin ; P20: output for pwm ; pwm_output.equ bitno0 P21-P27: output idle N.C. ; ; Port 3 pin P31: output voltage sensing ; Usense .equ bitnol P32: idle input, pull to ground when not used ; P33: voltage reference ; Uref .equ bitno3 REGISTERS DEFINITION integration_timer .equ r4 ; counts the number of periods duration_of_high_cycle .equ r7 ; length of high time of period .set %FE ; counter for delay routine ; delay_counter

BITS DEFINITION CONSTANTS DEFINITION prel_min .equ 01111011b ; PRE1=30, continuous mode, int. clock ; pwm freq. = ; osc. freq./(PRE1 * 8*no_voltage_levels) ; in this case pwm freq. = 521Hz ; pwm freq. = 8E6Hz/(30*8*64)no_voltage_levels .equ 64 ; number of voltage levels start_up_voltage .equ 2 ; voltage start level <=</pre> no_voltage_levels ; max duration of high cycle max_high_cycle .equ 63 255 ; extends start up time by 3 inv_softstart_time .equ ; extension factor = 765/inv_softstart_time MACROS ; Refer to Z8 technical manual for macro definition ; MACRO register, bitnumber ; set the appropriate bit in or \register, #\bitnumber ; the specified register bset MACEND MACRO register, bitnumber ; clear the appropriate bit in bclr \register,# ~(\bitnumber) ; the specified register and MACEND MACRO register, bitnumber, label ; IF the appropriate bit in brset tcm \register,#\bitnumber ; the specified register is set ; THEN jump to label ir z,\label MACEND ; ELSE go on MACRO register, bitnumber, label brclr ; IF the appropriate bit in the tm \register,#\bitnumber ; specified register is reset z,\label ; THEN jump to label jr MACEND ; ELSE go on pwm_high_cycle MACRO bset r2,pwm_output MACEND le MACRO bclr r2,pwm_output pwm_low_cycle MACEND INTERRUPTS VECTOR .MLIST .LIST ; Interrupt vector address %00 to %0C .ORG 80000 irq0 .word .word irql .word irq2 .word irq3 .word irq4 .word timer1 PROGRAM STARTS HERE ;

BEGINNING: .ORG %0C irq0: irq1: irq2: irq3: irq4: di ld P01M,#00000101b ;P0, P1 input, internal stack ld P2M,#0000000b ;P20-P27 output ld P3M,#00000011b ;P3 analog + P2 push pull ;Switch off transistor and P2,#11111110b clr SPH ld SPL,#%40 ; INIT STACK POINTER IPR,#00001010b ; IRQ5 has highest priority ld ld IMR,#00100000b ; enable T1 interrupt ; INITIALIZE RAM TO "0" #%30 srp ï ; ld R14,#%3d clr @R14 ;zram: djnz R14,zram ; Initialize all registers ; srp #%00 ; set working register to %00 clr integration timer ld prel, #prel_min ; preset T1 ld duration_of_high_cycle,#start_up_voltage ; preset voltage level T1,duration_of_high_cycle
duration_of_low_cycle,#no_voltage_levels ; ld ld ; duration_of_low_cycle,duration_of_high_cycle ; sub ei TMR,#00011100b ld SOFTSTART_WISHED IF call softstart ; increase voltage slowly ENDIF MAIN USER PROGRAM ; The step down voltage regulator runs as a batch task via T1 interrupt. ; Main: NOP ; insert your instructions ; here jp Main SUBROUTINES _____ IF SOFTSTART_WISHED $1.5uS \times (30x222) = 10 mS$; .set %FE delay_counter

delay10msec	c: ld	delay_counter,#222 ; 6 c	cycles		
loop1:	nop nop dec jr ret	delay_counter ; 6 c nz,loop1 ; 12 cycles	; 6 cycles ; 6 cycles cycles		
delay100ms6	call call call call call call call call	delay10msec delay10msec delay10msec delay10msec delay10msec delay10msec delay10msec delay10msec delay10msec			
ENDIF					
<pre>;////////////////////////////////////</pre>					
softstart_counter .set r5					
softstart:	ld	r3,Usense,end_softstart softstart_counter,#inv_softst delay100msec softstart_counter,softstart_c			
slow_down:	dec ;ld ;sub jr	<pre>duration_of_high_cycle duration_of_low_cycle,#no_vol duration_of_low_cycle,duratic softstart</pre>			
end_softstart: ret ENDIF					
<pre>;////////////////////////////////////</pre>					
timer1:	push srp	rp #%00	<pre>; working register ; group 0 reserved ; for timer1 ; interrupt</pre>		
turn off:	brclr	r2,pwm_output,turn_on	; If last pwm-cycle was ; low then next pwm- ; cycle is high		
turn oll.					

turn_on:	jr	end_of_interrupt	<pre>; low cycle is finished, and ; duration_of_high_cycle ; is loaded from T1 into ; 8bit-down-counter</pre>		
	ld	igh_cycle T1,duration_of_low_cycle ;	<pre>; On next T1 end_of_count, high cycle is finished, ; and duration_of_low_cycle ; is loaded from T1 into 8bit- ;</pre>		
down-counter					
<pre>;////////////////////////////////////</pre>					
end_of_interrupt: pop rp iret					
	END				

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